

POWER MANAGEMENT

Dual EL Lamp Driver

The IMP522 is a dual-output, high-voltage electroluminescent (EL) lamp driver. Either or both EL lamp driver outputs can be turned ON with the LMPSEL select pin. One EL lamp is connected between V_A and V_{AB} and the other is connected between V_B and V_{AB} . V_{AB} is a common pin for both lamps. With an input supply voltage between 2.0V and 6.5V, the typical regulated lamp drive voltage is 220V peak-to-peak.

The device uses a single inductor and a minimum number of passive components: a storage capacitor, a fast recovery diode and two resistors to set the PWM and EL drive frequencies. These can be independently set to optimize brightness and minimize power consumption. R_{SW} is connected between the $R_{SW\text{-}OSC}$ pin and the supply pin V_{DD} to set the frequency for the internal 4.0Ω switching MOSFET. The switch duty cycle is 88%. The EL lamp driver frequency is set by R_{EL} connected between the $R_{EL\text{-}OSC}$ pin and the V_{DD} pin.

Designed to minimize battery current drain, the IMP522 draws 2mA maximum. A power-saving shutdown mode reduces current to 2µA maximum.

The IMP522 is available in a compact 10-pin MicroSO package and in die form.

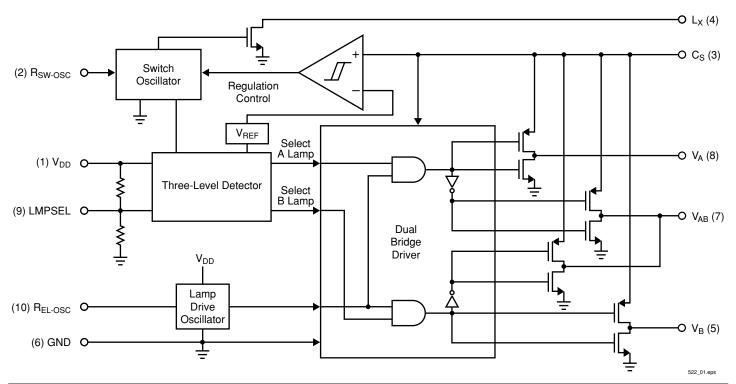
Key Features

- Drive two EL lamps independently
- ◆ Digital LMPSEL pin
 - Activate either or both EL output drivers
- ◆ 220V_{P-P} typical AC output voltage drives 30nF EL lamps
- ♦ Wide operating voltage range: 2V to 6.5V
- ◆ Low current consumption: 2mA maximum
- ◆ Disable mode extends battery life
 − Disable current 2µA maximum
- ◆ Compact 10-pin MicroSO package
- ◆ High-voltage, low-cost CMOS process

Applications

- ◆ Cellular phones
- ◆ PDAs/Handheld computers
- ◆ Toys/Consumer electronics
- ◆ Safety Illumination
- LCD modules
- Remote controls

Block Diagram





Pin Configuration

$\begin{array}{c|c} \textbf{MicroSO} \\ \textbf{V}_{DD} & \textbf{1} \\ \textbf{R}_{SW\text{-OSC}} & \textbf{2} \\ \textbf{C}_{S} & \textbf{3} \\ \textbf{L}_{X} & \textbf{4} \\ \textbf{V}_{B} & \textbf{5} \\ \end{array} \quad \begin{array}{c|c} \textbf{IMP522} & \textbf{10} & \textbf{R}_{EL\text{-OSC}} \\ \textbf{9} & \textbf{LMPSEL} \\ \textbf{8} & \textbf{V}_{A} \\ \textbf{7} & \textbf{V}_{AB} \\ \textbf{6} & \textbf{GND} \\ \end{array}$

Ordering Information

Part Number	Input Voltage	Temperature Range	Pins-Package
IMP522EMB	2.0V to 6.5V	−40°C to +85°C	10-MicroSO

Add /T to ordering part number for Tape and Reel.

Absolute Maximum Ratings

V_{DD} , R_{SW-OSC} and R_{EL-OSC}	0.5V to +7.0V
C_S , L_X	0.5V to +120V
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Power Dissipation (MicroSO)	500mW
V_A , V_{B} , V_{AB}	–0.5V to $V_{CS} (pin 3)$

Note: All voltages are referenced to GND.

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability.

Electrical Characteristics

Unless otherwise noted, V_{DD} = 3.0V, R_{SW} = 910k Ω , R_{EL} = 2.7M Ω , L = 220 μ H and T_A = 25°C.

Parameter	Symbol	Conditions	Min	Тур	Max	Units
ON-resistance of MOS Switch	R _{DS(ON)}	I = 100mA		4.0	8.0	Ω
Output Voltage Regulation	V _{CS}	$V_{DD} = 2.0 \text{ to } 6.5 \text{V},$ T = -40°C to 85°C		110	120	V
Output Voltage Peak-to-Peak (in regulation)	V_A - V_{AB} , V_B - V_{AB}	In Regulation		220		V
Input Current at V _{DD} Pin	I_{DD}	LMPSEL = GND		650		μΑ
Powerdown Input Current	I _{DDQ}	$V_{RSW-OSC}$ < 100mA V_{DD} = 2.0 to 6.5 T = -40°C to 85°C			2	μА
Input Crrent Plus Inductor Current	I _{IN}	See Figure 1 LMPSEL = GND or V _{DD} LMPSEL = floating		30 43		mA mA
Output Drive Frequency	f _{EL}	-		250		Hz
Switching Frequency	f _{SW}			61		kHz
Switching Duty Cycle	D _{SW}			88		%
LMPSEL Low-Level Threshold	V_{IL}			0.3 V _{DD}		V
LMPSEL High-Level Threshold	V _{IH}			0.7 V _{DD}		V
LMPSEL Input Resistance	R _{LMPSEL}			50		kΩ
LMPSEL Sink/Source Resistance	I _{LMPSEL}	Floating/High Impedance State	± 5			μΑ
LMPSEL Hysteresis	V_{hys}			50		mV



Pin Descriptions

Pin Number	Name	Function
1	V_{DD}	Positive voltage supply. Inductor L may be connected here or to a separate unregulated supply.
2	R _{SW-OSC}	Switch-mode resistor pin. The external resistor R _{SW} determines switching frequency.
3	Cs	Boost converter storage capacitor. The voltage across the EL lamp is approximately equal to twice the voltage at $C_{\rm S}$.
4	L _X	Connection to flyback inductor L.
5	V_{B}	Output for EL Lamp B.
6	GND	Ground.
7	V _{AB}	Common terminal for both EL lamps.
8	V _A	Output for EL Lamp A.
9	LMPSEL	Digital three-state input pin. Select either lamp A or lamp B or both lamps.
10	R _{EL-OSC}	The EL lamp oscillator frequency setting pin. External resistor R_{EL} connected to V_{DD} sets the EL Lamp drive frequency for both lamps.

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External Components

External Component	Description and Selection Guide
Diode	Catch diode. A fast reverse recovery diode, with BV > 150V, such as an FDLL400 (150V).
Capacitor C _S	This is the high voltage capacitor that stores the inductive energy transferred through the catch diode. A capacitor with WV > 120V between 10nF and 100nF is recommended.
Resistor R _{EL}	The EL lamp oscillator frequency setting resistor. This resistor, connected between the $R_{EL\text{-}OSC}$ pin and V_{DD} , provides an oscillator frequency inversely proportional to R_{EL} ; as R_{EL} increases, the EL lamp frequency decreases along with the current drawn by the lamp. Lamp color is also determined by this frequency. A 2.7M Ω resistor between the $R_{EL\text{-}OSC}$ pin and the V_{DD} supply results in a lamp frequency around 250Hz.
Resistor R _{SW}	Switching Oscillator frequency setting resistor. The switching oscillator resistor is connected between the $R_{SW\text{-}OSC}$ pin and the V_{DD} supply. The switching frequency is inversely proportional to the resistor value, dropping as the resistance increases.
Inductor L	The inductor provides the voltage boost needed by means of inductive "flyback". The internal MOSFET switch alternately opens and closes the ground connection for the inductor at the L_X pin. When this internal switch opens, the inductor potential will forward-bias the catch diode and the current will pass through the storage capacitor C_S , charging it to a high voltage.
	Smaller inductors are preferred to prevent saturation. As the value of the inductor increases (and the series DC resistance of the inductor decreases), the switching frequency set by R _{SW} should be increased to prevent saturation. In general, smaller value inductors that can handle more current are more desirable when larger area EL lamps must be driven.
Resistor R _{AB}	Lamp waveform shaping resistor for low audible noise applications.



General Purpose Application Circuit, 3.0V

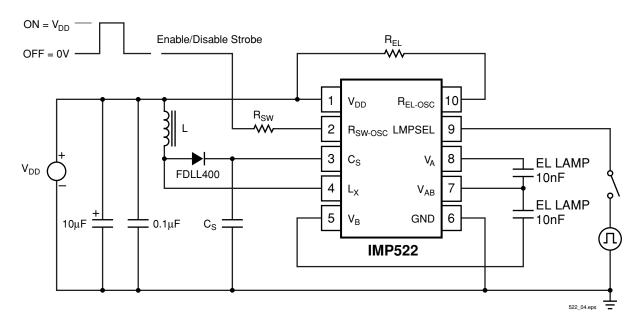


Figure 1. General Purpose Application

Table 1. Bill of Materials

Component	Value	Specifications	Make and Part Number
R _{SW}	910kΩ	0.1W	Any
R _{EL}	2.7ΜΩ	0.1W	Any
L	220μΗ	$I_{DC} = 64 \text{mA}/39 \Omega$	TOKO 875FU-222M
C _S	3.3nF	100V	NOVACAP 0805B332J101N

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Low Audible Noise for Mobile Phone Application Circuit

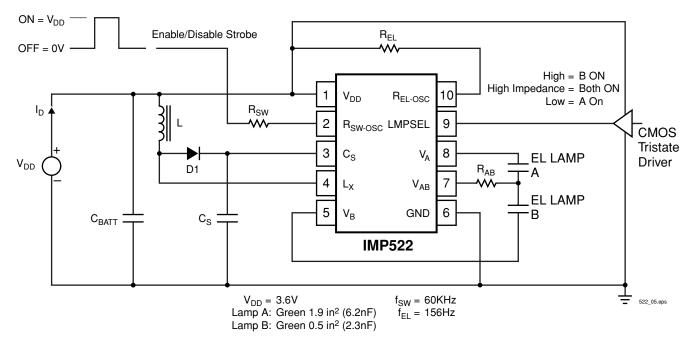


Figure 2. Low Audible Noise for Mobile Phone Application Circuit

Table 2. Bill of Materials

Component	Value	Specifications	Make and Part Number
R _{SW}	1.1ΜΩ	0.1W	Any
R _{EL}	4ΜΩ	0.1W	Any
R _{AB}	18K	0.1W	Any
C _{BATT}	0.1μF	10V	Any
C _S	3.3nF	100V	NOVACAP 0805B332J101N
L	2200μΗ	$I_{DC} = 64\text{mA}/39\Omega$	TOKO 875FU-222M
D1	_	150V/200mA	FDLL 400

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Application Information

EL Lamp Drive

The outputs $V_A - V_{AB}$ and $V_B - V_{AB}$ are configured as H-bridges, driven by the EL oscillator. Each output is switched between C_S and ground on alternate phases, creating peak-to-peak signals across the EL lamps of twice the regulated voltage.

EL Lamp Selection: LMPSEL

The digital input pin LMPSEL allows either or both EL lamps to be active. Lamp A is active when LMPSEL is LOW and lamp B is active when LMPSEL is HIGH. When LMPSEL is left floating or driven by a three-state driver in the high impedance state, both lamp driver outputs are active.

LMPSEL Signal	Lamp A Drive V _A and V _{AB}	Lamp B Drive V _B and V _{AB}
HIGH	OFF	ON
LOW	ON	OFF
Floating/ High Impedance	ON	ON

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The logic HIGH signal level is defined as greater than $0.7V_{DD}$ and logic LOW is defined as less than $0.3V_{DD}$. A floating level is recognized with the signal level between $0.3V_{DD}$ and $0.7V_{DD}$, or when the output impedance of the driving voltage signal source is infinite (driver in OFF state).

Both drivers are OFF if the IMP522 is disabled.

EL Driver Output Overvoltage Regulator

The IMP522 maximum V_{CS} output voltage is between 110V and 120V. The internal overvoltage regulator skips the inductor switching whenever the voltage on the C_S pin exceeds the regulation threshold. The internal overvoltage detection trip point has a hysterisis of 1V and a range of 110V to 120V at room temperature.

PWM Circuit Switching

The switching MOSFET is driven by the PWM signal (nominally 61kHz). During the first 88% of the period, the switch is ON, providing a low impedance path ($<8\Omega$) from L_X to ground. This causes the external inductor to charge. In the last 12% of the period, the MOSFET is turned OFF. This causes the voltage on the output of L_X to rise up to a high value. At some point, this will forward-bias the external diode, thus pumping charge into the storage capacitor C_S . The voltage on C_S increases each cycle to between 110V and 120V. When the internal regulation trip-point is reached, the overvoltage regulator turns the MOSFET switch OFF to conserve power.

Power Sequencing

To power up the chip, the $R_{SW\text{-}OSC}$ pin is connected to V_{DD} through the external R_{SW} resistor. The voltage on the pin will charge up to $V_{DD}/2$. An internal threshold detector circuit monitors the pin voltage and when it exceeds the threshold range (0.2V to 0.9V) it powers up the oscillator and internal bias modules. This starts a delay counter which is one half of the EL oscillator period, after which power to the high voltage internal modules is applied. The IMP522 is then operating fully.

To power down the chip, R_{SW} is driven to ground via a switch or logic gate. When the voltage on the driver side of the resistor falls below $V_{DD}/2$, there will be no input bias current into the R_{SW-OSC} pin. This immediately powers down the internal high-voltage circuits, which effectively shuts the lamp off. At this point the oscillator and bias modules still draw quiescent current, but oscillations have ceased. As the R_{SW-OSC} pin voltage falls below 0.1, the oscillator and bias modules are also fully powered down.



Application Information

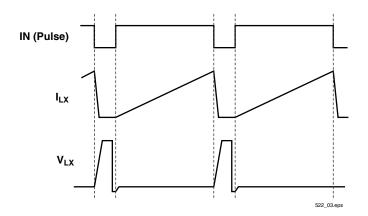


Figure 3. Driver Waveforms

Power Saving Disable Mode

The IMP522 can be powered up and down with $R_{\text{SW-OSC}}$. In normal operation, this resistor on the $R_{\text{SW-OSC}}$ pin is connected to V_{DD} or another voltage source. To power down (disable) the IMP522, R_{SW} is connected to ground.

When disabled, the IMP522 quiescent current drops to typically 20nA.

In die form, an extra pin ENABLE is available (contact factory). Connecting this pad to V_{DD} disables the chip. The ENABLE signal can be driven by a microcontroller.

Oscillator Frequency Adjustment

The EL lamp drive and PWM boost converter oscillation frequencies can be programmed independently.

The R_{SW} resistor, connected between the $R_{SW\text{-}OSC}$ pin and V_{DD} , determines the Inductor Switching (or PWM-) frequency. For the recommended nominal resistor value of $910k\Omega$, the frequency is 61kHz. For other resistor values, the frequency is inversely proportional to the resistor value. Increasing the resistance will lower the frequency.

The R_{EL} resistor, connected between the $R_{EL\text{-OSC}}$ pin and V_{DD} , determines the EL lamp drive frequency. For the recommended nominal resistor value of 2.7M Ω , the frequency is 250Hz. For other resistor values, the frequency is inversely proportional to the resistor value: increasing the resistance will lower the frequency.

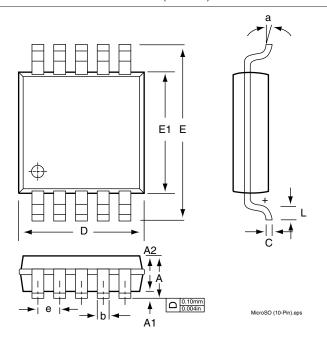
Oscillator	Nominal Resistor	Nominal Frequency
EL Lamp Drive	$R_{EL} = 2.7 M\Omega$	250Hz
Inductor Switch (PWM)	$R_{SW} = 910k\Omega$	61kHz

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Package Dimensions

MicroSO (10-Pin)



Inches			Millin	neters
	Min Max		Min	Max
		MicroSO (1	0-Pin)	
Α		0.0433		1.10
A1	0.0020	0.0059	0.050	0.15
A2	0.0307	0.0370	0.78	0.94
b	0.0059	0.0098	0.15	0.25
С	0.0051	0.0091	0.13	0.23
D	0.1142	0.1220	2.90	3.10
е	e 0.0197 BSC		0.50	BSC
Е	0.1990	0.1871	5.05	5.05
E1	0.1142	0.1220	2.90	3.10
L	0.0157	0.0276	0.40	0.70
а	0°	6°	0°	6°

Thermal Resistance		
θ _{JC}	41°C/W	
θ_{JA}	113°C/W	
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Daily Silver IMP Microelectronics Co.,Ltd 7 keda Road , Hi-Tech Park,

NingBo,Zhejiang, P.R.C. Post Code: 315040

Tel:(086)-574-87906358 Fax:(086)-574-87908866 e-mail:sales@ds-imp.com.cn Website:www.ds-imp.com.cn